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Features

- Supply-voltage Range: 2.7 V to 5.5 V
- Single-ended Output, no Balun Required
- Single-ended Input for RF and LO
- Excellent Isolation Characteristics
- Power-down Mode
- IP3 and Compression Point Programmable
- 2.5-GHz Operating Frequency

Benefits

- Reduced System Costs as only Few External Component (no Balun) are Required
- Small Package
- Very Low Current Consumption
- Easy to Use

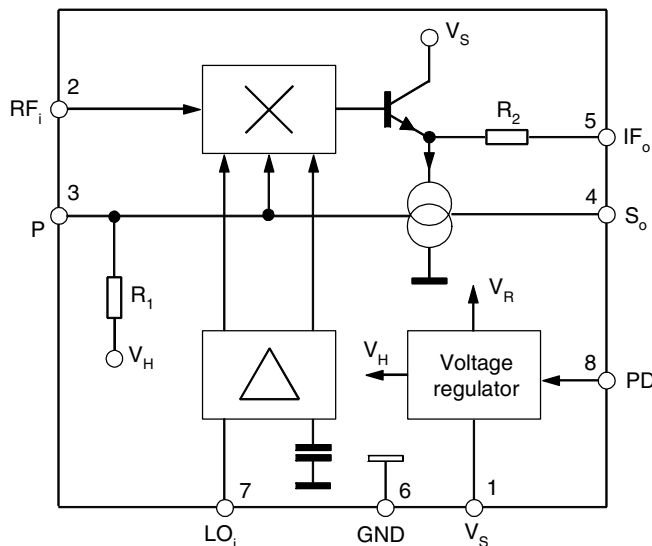
Electrostatic sensitive device.
Observe precautions for handling.



Description

The U2795B is a 2.5-GHz mixer for WLAN and RF telecommunications equipment, e.g., DECT and PCN. The IC is manufactured using Atmel's advanced bipolar technology. A double-balanced approach was chosen to assure good isolation characteristics and a minimum of spurious products. The input and output are single-ended, and their characteristics are programmable. No output transformer or balun is required.

Figure 1. Block Diagram

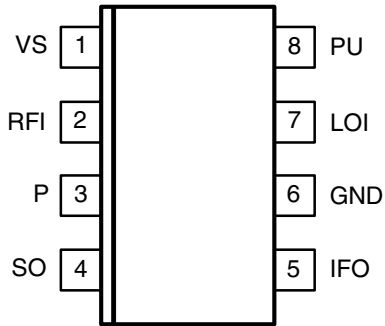


2.5-GHz Double- balanced Mixer

U2795B

Pin Configuration

Figure 2. Pinning



Pin Description

Pin	Symbol	Function
1	VS	Supply voltage
2	RFI	RF input
3	P	Programming port IP3, CP
4	SO	Output symmetry
5	IFO	IF output
6	GND	Ground
7	LOI	LO input
8	PU	Power-up

Functional Description

- Supply Voltage** The IC is designed for a supply-voltage range of 2.7 V to 5.5 V. As the IC is internally stabilized, the performance of the circuit is nearly independent of the supply voltage.
- Input Impedance** The input impedance, Z_{RFI} , is about 700 Ω with an additional capacitive component. This condition provides the best noise figure in combination with a matching network.
- 3rd Order Intercept Point (IP3)** The voltage divider, R_p/R_1 , determinates both the input and output intercept point, IIP3 and OIP3. If the value of R_p is infinite, the maximum value of IIP3 reach about -4 dBm. The IP3/ R_p characteristics are shown in Figure 3 and Figure 4.
- Output Impedance and Intercept Point** The output impedance is shown in Figure 11 on page 8. Both low output impedance and a high intercept point are defined to a high value of R_p .
- Current Consumption, I_S** Depending on the chosen input and output conditions of the IC, the current consumption, I_S , is between 4 mA and 10 mA. The current consumption in dependence of R_p is shown in Figure 6 on page 6.
- Power-up** This feature provides extended battery lifetime. If this function is not used, pin 8 has to be connected to V_S (pin 1).
- Output Symmetry** The symmetry of the load current can be matched and thus optimized for a given load impedance.

Absolute Maximum Ratings

Stresses beyond those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Parameters	Symbol	Value	Unit
Supply voltage	V_S	6	V
Input voltage	V_I	0 to V_S	V
Junction temperature	T_j	125	°C
Storage-temperature range	T_{stg}	-40 to +125	°C

Thermal Resistance

Parameters	Symbol	Value	Unit
Junction ambient SO8	R_{thJA}	175	K/W

Operating Range

Parameters	Symbol	Value	Unit
Supply-voltage range	V_S	2.7 to 5.5	V
Ambient-temperature range	T_{amb}	-40 to +85	°C

Electrical Characteristics

$V_S = 3\text{ V}$, $f_{LOi} = 1\text{ GHz}$, $IF = 900\text{ MHz}$, $RF = 100\text{ MHz}$, $R_P = \infty$ system impedance $Z_o = 50\ \Omega$, $T_{amb} = 25^\circ\text{C}$, $R_T = 56\ \Omega$
reference point pin 6, unless otherwise specified

No.	Parameters	Test Conditions	Pin	Symbol	Min.	Typ.	Max.	Unit	Type*
1.1	Supply voltage range		1	V_S	2.7		5.5	V	A
1.2	Supply Current	$V_S = 2.7\text{ V}$, $R_P = 10\text{ k}\Omega$	1	I_S	9		13	mA	A
		$V_S = 3.0\text{ V}$, $R_P = \infty$	1	I_S	3		6.2	mA	A
1.3	Conversion Power Gain	$R_L = 50\ \Omega$, $R_T = \infty$	1	PG_C		9		dB	B
		$R_L = 50\ \Omega$, $R_T = 56\ \Omega$		PG_C		4		dB	
2	Operating Frequencies								
2.1	RF_i frequency		2	RF_i	10		2500	MHz	D
2.2	LO_i frequency		7	f_{LOi}	50		2500	MHz	D
2.3	IF_o frequency		5	f_{IFo}	50		2500	MHz	D
3	Isolation								
3.1	LO spurious at RF_i	$P_{iLO} = -10\text{ to }0\text{ dBm}$	7, 2	IS_{LO-RF}		-30		dBm	D
3.2	RF_i to LO_i	$P_{iRF} = -25\text{ dBm}$	2, 7	IS_{RF-LO}		35		dB	D
3.3	LO spurious at IF_o	$P_{iLO} = -10\text{ to }0\text{ dBm}$	5, 7	IS_{LO-IF}		-25		dBm	D
3.4	IF_o to LO_i		5, 7	IS_{IF-LO}		30		dB	D
4	Output (IF)								
4.1	Output compression point		5	CP_O		-10		dBm	D
5	Input (RF)								
5.1	Input impedance		2	Z_{RFi}		$700\ \Omega \parallel 0.8$		$\Omega \parallel pF$	D
5.2	Input compression point		2	CP_i		-14		dBm	D
5.3	3rd-order input intercept point		2	IIP3		-4		dBm	D
6	Input (LO)								
6.1	LO level		7	P_{iLO}		-6		dBm	D
7	Voltage Standing Wave Ratio (VSWR)								
7.1	Input LO		7	$VSWR_{LOi}$		< 2			D
7.2	Output IF		4	$VSWR_{IFo}$		< 2			D
8	Noise Performance								
8.1	Noise figure	$P_{iLO} = 0\text{ dBm}$, $R_T = \infty$		NF		10		dB	D
9	Power-down Mode								
9.1	Supply current	$V_{PU} < 0.5\text{ V}$	1	I_{SPU}		< 5	30	μA	B
		$V_{PU} = 0\text{ V}$						μA	B
10	Power-down Voltage								
10.1	"Power ON"	$V_S = 3.5\text{ to }5.5\text{ V}$	8	V_{PON}	$V_S - 0.5$		$V_S + 0.5$	V	D
		$V_S = 2.7\text{ to }3.5\text{ V}$			V_S		$V_S + 0.5$	V	D
10.2	"Power DOWN"		8	V_{PDN}			1	V	D
10.3	Power-down current	Power ON	8	I_{PON}		0.15	0.22	mA	A
		Power DOWN		I_{PDN}		< 5		μA	D
10.4	Settling time		5,8	t_{SPD}		< 30		μs	D

*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Figure 3. IIP3 versus Resistor R_p , IF: 900 MHz

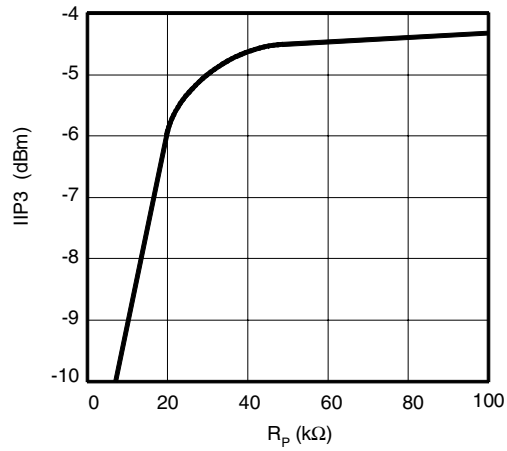


Figure 4. OIP3 versus Resistor R_p , IF: 900 MHz

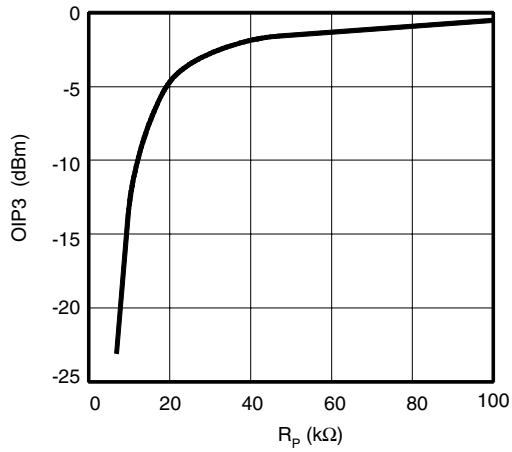


Figure 5. Gain versus Resistor R_p , LO: 1030 MHz, level -10 dBm; RF: 130 MHz, -30 dBm, $R_T = 56 \Omega$

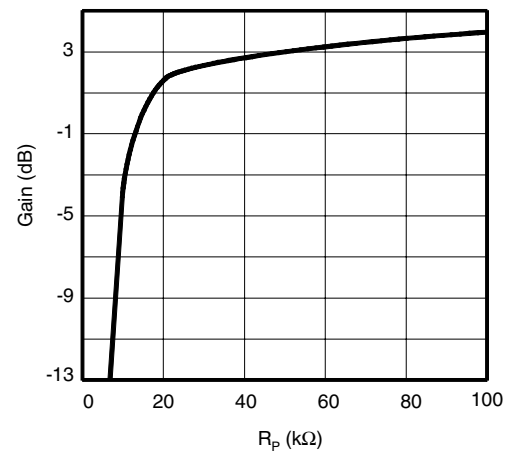


Figure 6. Supply Current I_S versus Resistor R_p

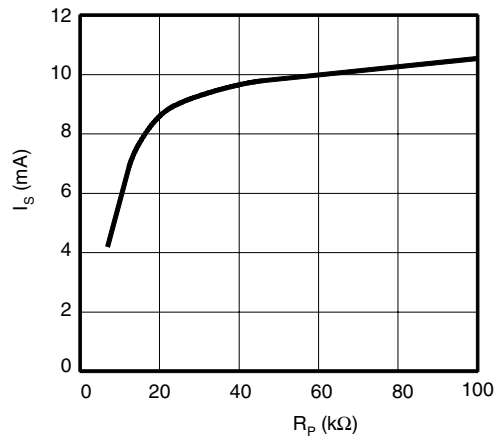


Figure 7. Gain versus IF Output Frequency, LO Level: -6 dBm, RF: 130 MHz, -35 dBm; Parameter: RF Input Termination

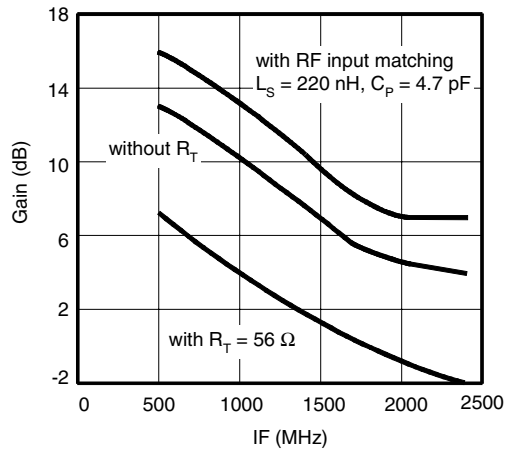


Figure 8. IIP3 versus IF Output Frequency, LO Level: -6 dBm; RF: 130 MHz/130.1 MHz, -35 dBm; Parameter: RF Input Termination

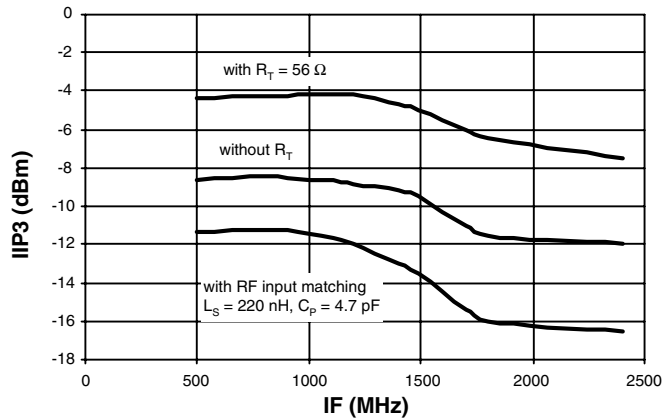


Figure 9. Double Sideband Noise Figure versus IF Output Frequency; LO: 1000 MHz, Level 0 dBm; no RF Input Matching, R_T Left Out

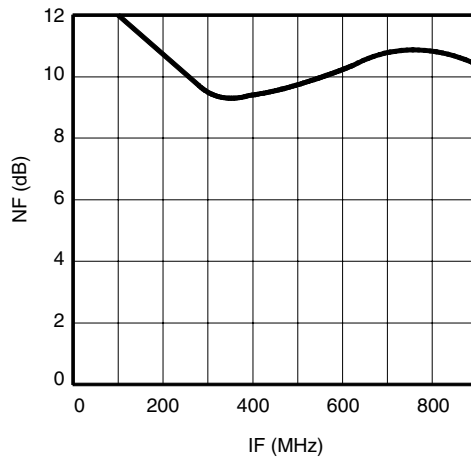


Figure 10. Typical VSWR Frequency Response of the IF Output, $R_p = \infty$

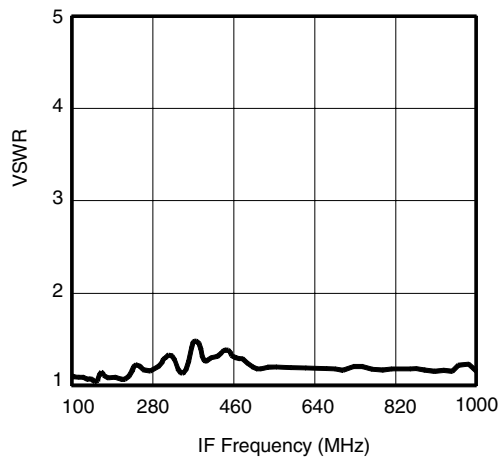


Figure 11. Typical Impedance of the Output versus R_p at Frequency $f_{IF0} = 900$ MHz
 Markers (from Left to Right): $R_p = \infty/22\text{ k}\Omega/10\text{ k}\Omega/8.2\text{ k}\Omega/5.6\text{ k}\Omega$

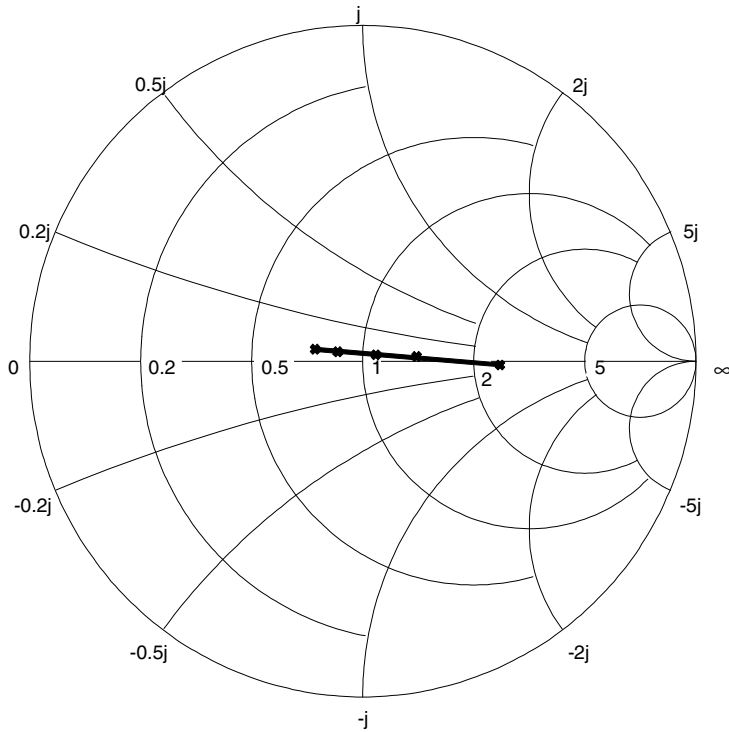


Figure 12. Typical S11 Frequency Response of the IF Output, $R_p = \infty$ IF Frequency from 100 MHz to 1000 MHz, Marker: 900 MHz

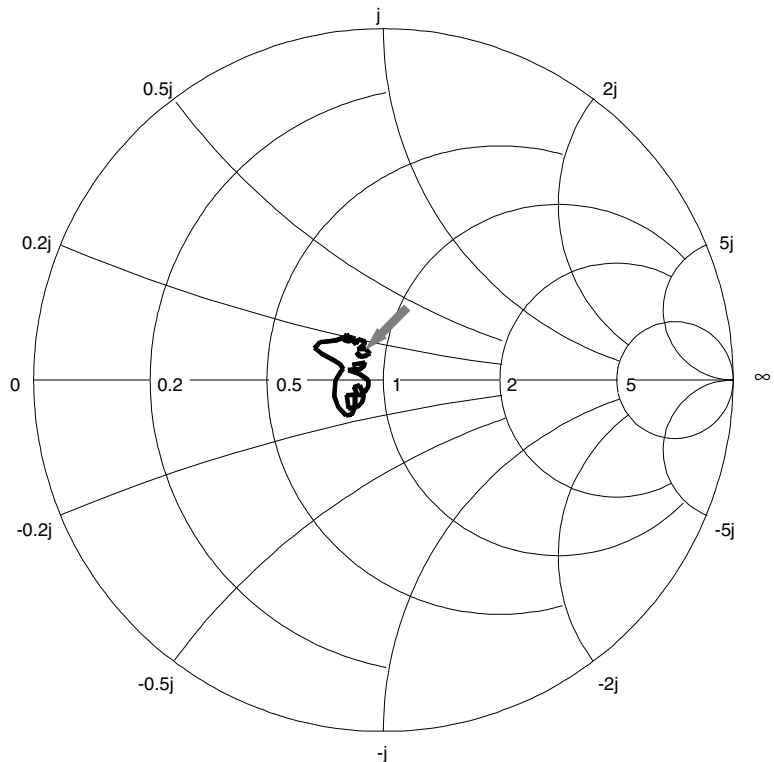


Figure 13. Typical S11 Frequency Response of the RF Input, $R_p = \infty$ $R_T = \infty$
 RF Frequency from 100 MHz to 1000 MHz, Marker: 900 MHz

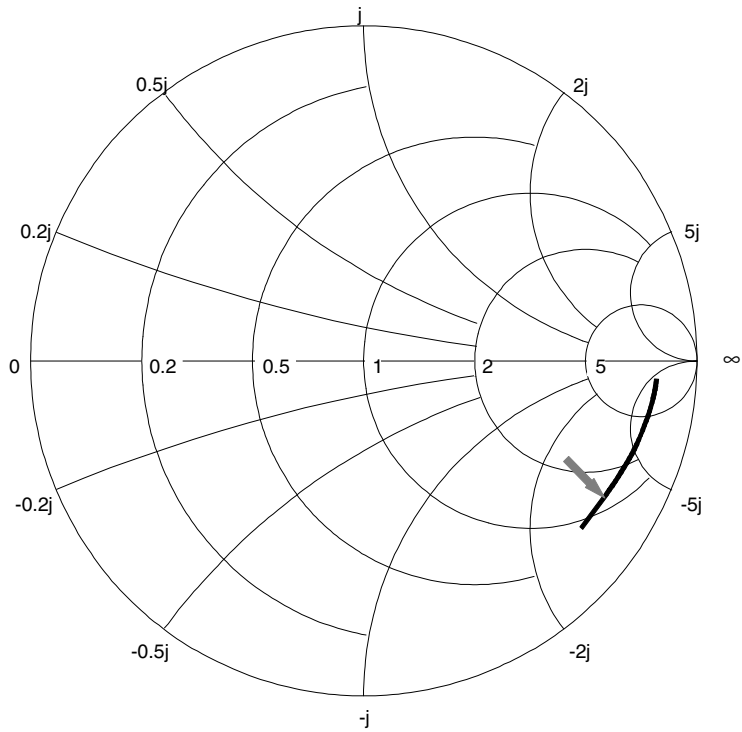
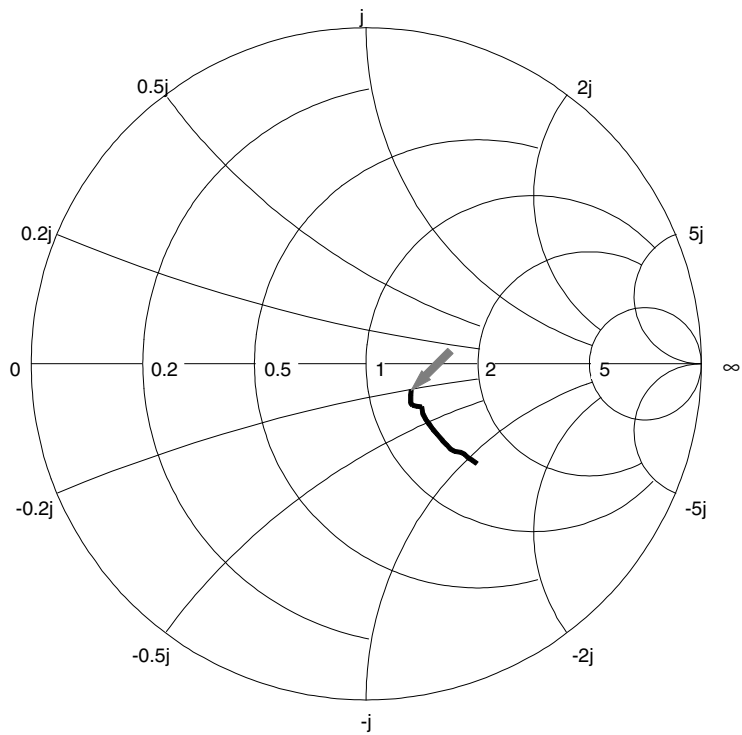


Figure 14. Typical S11 Frequency Response of the LO Input, $R_p = \infty$ LO Frequency
 from 100 MHz to 1000 MHz, Marker: 900 MHz



Application

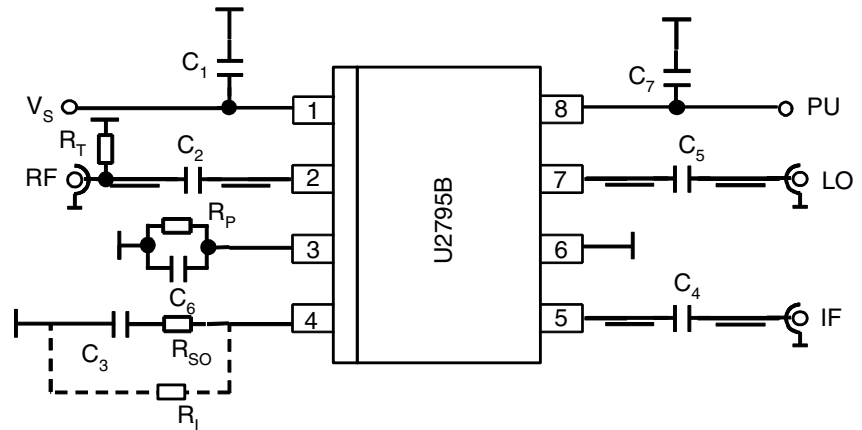


Table 1. Part List

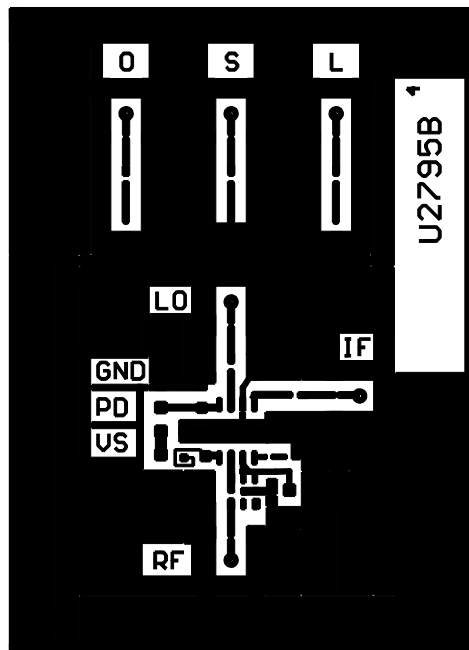
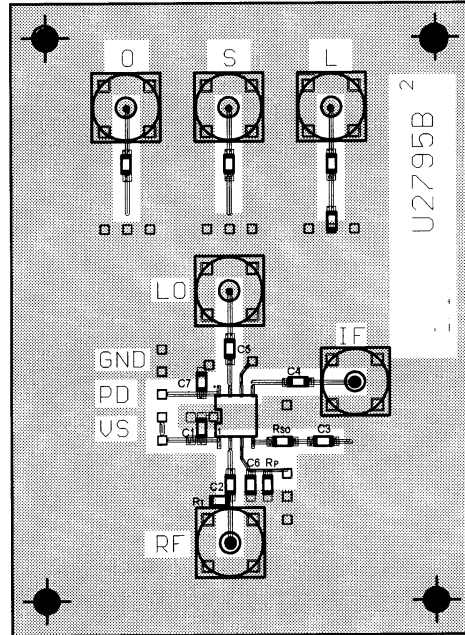
Part	Value
C_1	10 nF
$C_2, C_3, C_4, C_5, C_6, C_7$	100 pF
$*R_P$	
	50-Ω Microstrip
$*R_{SO}$	68 Ω
	optional
R_T	56 Ω

If the part-list values are used, the PU settling time is $< 20 \mu\text{s}$. Using other values, time requirements in burst-mode applications have to be considered.

The values of R_{SO} and R_P depend on the input and output condition requirements. For R_{SO} , 68 Ω is recommended.

By means of the optional R_I , the intercept and compression point can be slightly increased; values between 500 Ω and 1 kΩ are suitable. Please note that such modification will also increase the supply current.

Application Circuit (Evaluation Board)





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