阅读申明

1.本站收集的数据手册和产品资料都来自互联网，版权归原作者所有。如读者和版权方有任何异议请及时告之，我们将妥善解决。
2.本站提供的中文数据手册是英文数据手册的中文翻译，其目的是协助用户阅读，该译文无法自动跟随原稿更新，同时也可能存在翻译上的不当。建议读者以英文原稿为参考以便获得更精准的信息。
3.本站提供的产品资料，来自厂商的技术支持或者使用者的心得体会等，其内容可能存在描叙上的差异，建议读者做出适当判断。
4.如需与我们联系，请发邮件到marketing@iczoom.com，主题请标有"数据手册"字样。

Read Statement

1. The datasheets and other product information on the site are all from network reference or other public materials, and the copyright belongs to the original author and original published source. If readers and copyright owners have any objections, please contact us and we will deal with it in a timely manner.

2. The Chinese datasheets provided on the website is a Chinese translation of the English datasheets. Its purpose is for reader's learning exchange only and do not involve commercial purposes. The translation cannot be automatically updated with the original manuscript, and there may also be improper translations. Readers are advised to use the English manuscript as a reference for more accurate information.

3. All product information provided on the website refer to solutions from manufacturers' technical support or users the contents may have differences in description, and readers are advised to take the original article as the standard.

4. If you have any questions, please contact us at marketing@iczoom.com and mark the subject with "Datasheets".

General Description

The 8534-01 is a low skew, 1-to-22 Differential-to-3.3V LVPECL Fanout Buffer. The 8534-01 has two selectable clock inputs. The CLK, nCLK pair can accept most standard differential input levels. The PCLK, nPCLK pair can accept LVPECL, CML, or SSTL input levels. The device is internally synchronized to eliminate runt pulses on the outputs during asynchronous assertion/deassertion of the OE pin. The 8534-01's low output and part-to-part skew characteristics make it ideal for workstation, server, and other high performance clock distribution applications.

Features

- Twenty-two differential LVPECL outputs
- Selectable differential CLK, nCLK or LVPECL clock inputs
- CLK, nCLK pair can accept the following differential input levels: LVPECL, LVDS, LVHSTL, HCSL, SSTL
- PCLK, nPCLK supports the following input levels: LVPECL, CML, SSTL
- Maximum output frequency: 500MHz
- Output skew: 100ps (maximum)
- Translates any single-ended input signal (LVCMOS, LVTTL, GTL) to LVPECL levels with resistor bias on nCLK input
- Additive phase jitter, RMS): 0.15ps (typical)
- Full 3.3V supply mode
- 0°C to 85°C ambient operating temperature
- Available in lead-free (RoHS 6) package

Block Diagram
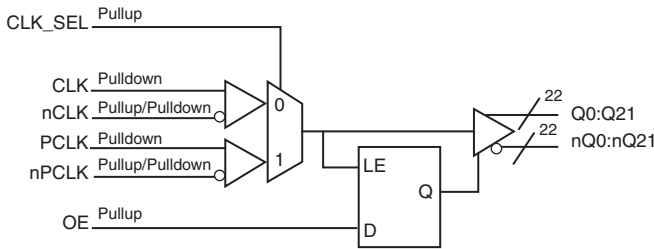
Pin Assignment
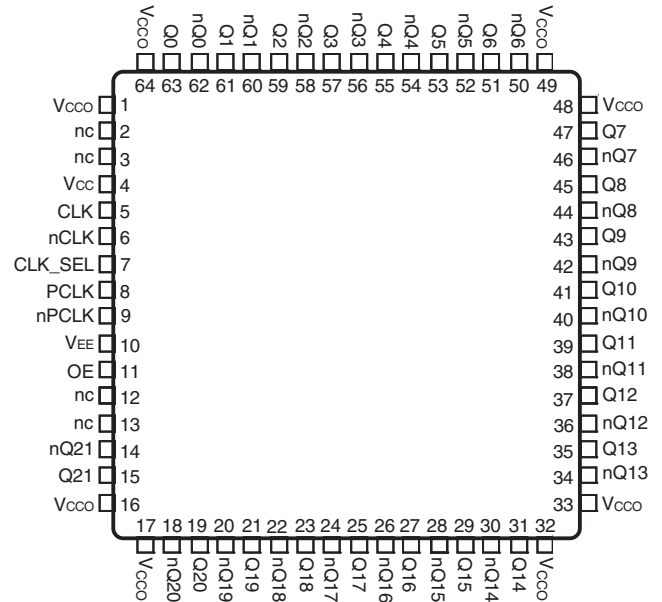
8534-01
64-Lead TQFP E-Pad
10mm x 10mm x 1.0mm package body
Y package
Top View

Pin Descriptions and Pin Characteristic Tables

Table 1. Pin Descriptions

Number	Name	Type		Description
1, 16, 17, 32, 33, 48, 49, 64	V_{CCO}	Power		Output supply pins for LVPECL outputs.
2, 3, 12, 13	nc	Unused		No connect.
4	V_{CC}	Power		Core supply pin for LVPECL outputs.
5	CLK	Input	Pulldown	Non-inverting differential clock input.
6	nCLK	Input	Pullup/ Pulldown	Inverting differential clock input. Pulled to $^2/_3$ V_{CC}.
7	CLK_SEL	Input	Pullup	Clock select input. When HIGH, selects PCLK, nPCLK inputs. When LOW, selects CLK, nCLK inputs. LVCMOS / LVTTL interface levels.
8	PCLK	Input	Pulldown	Non-inverting differential LVPECL clock input.
9	nPCLK	Input	Pullup/ Pulldown	Inverting differential LVPECL clock input. Pulled to $^2/_3$ V_{CC}.
10	V_{EE}	Power		Negative supply pin.
11	OE	Input	Pullup	Output enable. When logic HIGH, the outputs are enabled (default). When logic LOW, the outputs are disabled and drive differential low: Qx = LOW, nQx = HIGH. LVCMOS / LVTTL interface levels.
14, 15	nQ21, Q21	Output		Differential clock outputs. LVPECL interface Levels.
18, 19	nQ20, Q20	Output		Differential clock outputs. LVPECL interface Levels.
20, 21	nQ19, Q19	Output		Differential clock outputs. LVPECL interface Levels.
22, 23	nQ18, Q18	Output		Differential clock outputs. LVPECL interface Levels.
24, 25	nQ17, Q17	Output		Differential clock outputs. LVPECL interface Levels.
26, 27	nQ16, Q16	Output		Differential clock outputs. LVPECL interface Levels.
28, 29	nQ15, Q15	Output		Differential clock outputs. LVPECL interface Levels.
30, 31	nQ14, Q14	Output		Differential clock outputs. LVPECL interface Levels.
34, 35	nQ13, Q13	Output		Differential clock outputs. LVPECL interface Levels.
36, 37	nQ12, Q12	Output		Differential clock outputs. LVPECL interface Levels.
38, 39	nQ11, Q11	Output		Differential clock outputs. LVPECL interface Levels.
40, 41	nQ10, Q10	Output		Differential clock outputs. LVPECL interface Levels.
42, 43	nQ9, Q9	Output		Differential clock outputs. LVPECL interface Levels.
44, 45	nQ8, Q8	Output		Differential clock outputs. LVPECL interface Levels.
46, 47	nQ7, Q7	Output		Differential clock outputs. LVPECL interface Levels.
50, 51	nQ6, Q6	Output		Differential clock outputs. LVPECL interface Levels.
52, 53	nQ5, Q5	Output		Differential clock outputs. LVPECL interface Levels.
54, 55	nQ4, Q4	Output		Differential clock outputs. LVPECL interface Levels.
56, 57	nQ3, Q3	Output		Differential clock outputs. LVPECL interface Levels.
58, 59	nQ2, Q2	Output		Differential clock outputs. LVPECL interface Levels.
60, 61	nQ1, Q1	Output		Differential clock outputs. LVPECL interface Levels.
62, 63	nQ0, Q0	Output		Differential clock outputs. LVPECL interface Levels.

NOTE: *Pullup and Pulldown* refer to internal input resistors. See Table 2, *Pin Characteristics,* for typical values.
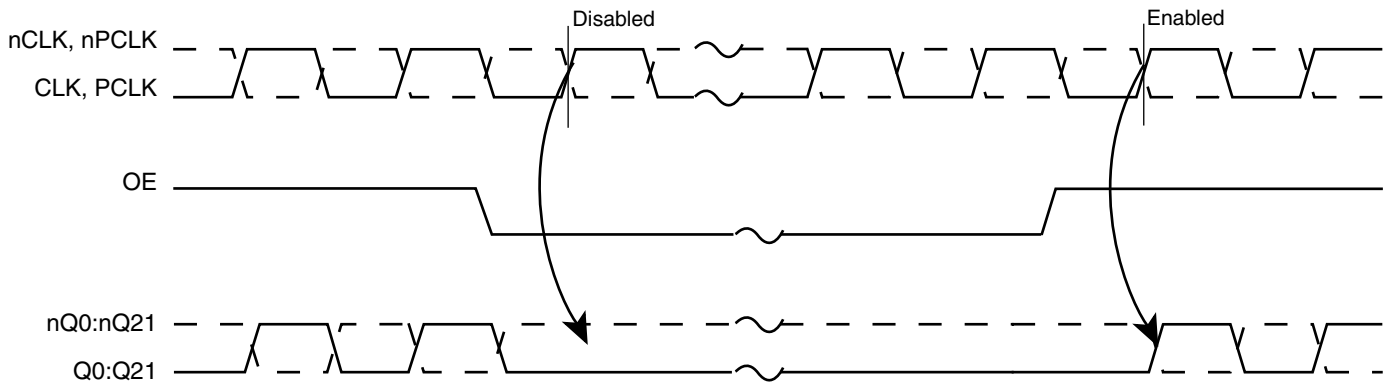
Table 2. Pin Characteristics

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
C_{IN}	Input Capacitance			4		pF
R_{PULLUP}	Input Pullup Resistor			37		kΩ
$R_{PULLDOWN}$	Input Pulldown Resistor			75		kΩ

Function Table

Table 3. Control Input Function Table.

Inputs		Outputs	
OE	CLK_SEL	Q0:Q21	nQ0:nQ21
0	0	LOW	HIGH
0	1	LOW	HIGH
1	0	CLK	nCLK
1	1	PCLK	nPCLK

Figure 1. OE Timing Diagram

Absolute Maximum Ratings

NOTE: Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device.
These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond
those listed in the *DC Characteristics or AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for
extended periods may affect product reliability.

Item	Rating
Supply Voltage, V_{CC}	4.6V
Inputs, V_I	-0.5V to V_{CC} + 0.5V
Outputs, I_O (LVPECL) Continuous Current Surge Current	 50mA 100mA
Package Thermal Impedance, θ_{JA}	22.3°C/W (0 lfpm)
Storage Temperature, T_{STG}	-65°C to 150°C

DC Electrical Characteristics

Table 4A. Power Supply DC Characteristics, V_{CC} = V_{CCO} = 3.3V ± 5%, V_{EE} = 0V, T_A = 0°C to 85°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V_{CC}	Core Supply Voltage		3.135	3.3	3.465	V
V_{CCO}	Output Supply Voltage		3.135	3.3	3.465	V
I_{EE}	Power Supply Current				230	mA

Table 4B. LVCMOS/LVTTL DC Characteristics, V_{CC} = V_{CCO} = 3.3V ± 5%, V_{EE} = 0V, T_A = 0°C to 85°C

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
V_{IH}	Input High Voltage			2		V_{CC} + 0.3	V
V_{IL}	Input Low Voltage			-0.3		0.8	V
I_{IH}	Input High Current	OE, CLK_SEL	V_{CC} = V_{IN} = 3.465V			5	µA
I_{IL}	Input Low Current	OE, CLK_SEL	V_{CC} = 3.465V, V_{IN} = 0V	-150			µA

Table 4C. Differential DC Characteristics, V_{CC} = V_{CCO} = 3.3V ± 5%, V_{EE} = 0V, T_A = 0°C to 85°C

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
I_{IH}	Input High Current	CLK	V_{CC} = V_{IN} = 3.465V			150	µA
		nCLK	V_{CC} = V_{IN} = 3.465V			5	µA
I_{IL}	Input Low Current	CLK	V_{CC} = 3.465V, V_{IN} = 0V	-5			µA
		nCLK	V_{CC} = 3.465V, V_{IN} = 0V	-150			µA
V_{PP}	Peak-to-Peak Voltage; NOTE 1			0.15		1.3	V
V_{CMR}	Common Mode Input Voltage; NOTE 1, 2			V_{EE} + 0.5		V_{CC} − 0.85	V

NOTE 1: V_{IL} should not be less than -0.3V.
NOTE 2: Common mode input voltage is defined as V_{IH}.

Table 4C. LVPECL DC Characteristics, V_{CC} = V_{CCO} = 3.3V ± 5%, V_{EE} = 0V, T_A = 0°C to 85°C

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
I_{IH}	Input High Current	PCLK	V_{CC} = V_{IN} = 3.465V			150	µA
		nPCLK	V_{CC} = V_{IN} = 3.465V			5	µA
I_{IL}	Input Low Current	PCLK	V_{CC} = 3.465V, V_{IN} = 0V	-5			µA
		nPCLK	V_{CC} = 3.465V, V_{IN} = 0V	-150			µA
V_{PP}	Peak-to-Peak Voltage			0.3		1.0	V
V_{CMR}	Common Mode Input Voltage; NOTE 1			V_{EE} + 1.5		V_{CC}	V
V_{OH}	Output High Voltage; NOTE 2			V_{CCO} − 1.4		V_{CCO} − 0.9	V
V_{OL}	Output Low Voltage; NOTE 2			V_{CCO} − 2.0		V_{CCO} − 1.7	V
V_{SWING}	Peak-to-Peak Output Voltage Swing			0.6		1.0	V

NOTE 1: Common mode input voltage is defined as V_{IH}.
NOTE 2: Outputs terminated with 50Ω to V_{CCO} − 2V.

AC Electrical Characteristics

Table 5. V$_{CC}$ = V$_{CCO}$ = 3.3V ± 5%, V$_{EE}$ = 0V, T$_A$ = 0°C to 85°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
f$_{OUT}$	Output Frequency				500	MHz
t$_{PD}$	Propagation Delay; NOTE 1	$f \leq 500$MHz	2.0		3.0	ns
$t$sk(o)	Output Skew; NOTE 2, 3				100	ps
$t$sk(pp)	Part-to-Part Skew; NOTE 3, 4				700	ps
tjit	Buffer Additive Phase Jitter, RMS, refer to Additive Phase Jitter section; NOTE 5	Integration Range: 12kHz - 20MHz		0.15		ps
t$_R$ / t$_F$	Output Rise/ Fall Time	20% to 80%	200		700	ps
t$_S$	Setup Time		1			ns
t$_H$	Hold Time		0.5			ns
odc	Output Duty Cycle	$f \leq 266$MHz	48		52	%
		$266 < f \leq 500$MHz	46		54	%

NOTE: All parameters measured at f$_{OUT}$ unless noted otherwise.
NOTE: Special thermal considerations may be required. See Applications Section.
NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. Device will meet specifications after thermal equilibrium has been reached under these conditions.
NOTE 1: Measured from the differential input crossing point to the differential output crossing point.
NOTE 2: Defined as skew between outputs at the same supply voltage and with equal load conditions.
NOTE 3: This parameter is defined in accordance with JEDEC Standard 65. Measured at the output differential cross points.
NOTE 4: Defined as skew between outputs on different devices operating at the same supply voltages and with equal load conditions at the same temperature. Using the same type of inputs on each device, the outputs are measured at the differential cross points.
NOTE 5: Driving only one input clock.

Additive Phase Jitter

The spectral purity in a band at a specific offset from the fundamental compared to the power of the fundamental is called the **dBc Phase Noise.** This value is normally expressed using a Phase noise plot and is most often the specified plot in many applications. Phase noise is defined as the ratio of the noise power present in a 1Hz band at a specified offset from the fundamental frequency to the power value of the fundamental. This ratio is expressed in decibels (dBm) or a ratio of the power in the 1Hz band to the power in the fundamental. When the required offset is specified, the phase noise is called a **dBc** value, which simply means dBm at a specified offset from the fundamental. By investigating jitter in the frequency domain, we get a better understanding of its effects on the desired application over the entire time record of the signal. It is mathematically possible to calculate an expected bit error rate given a phase noise plot.

Additive Phase Jitter, RMS @ 156.25MHz
12kHz to 20MHz = 0.15ps (typical)
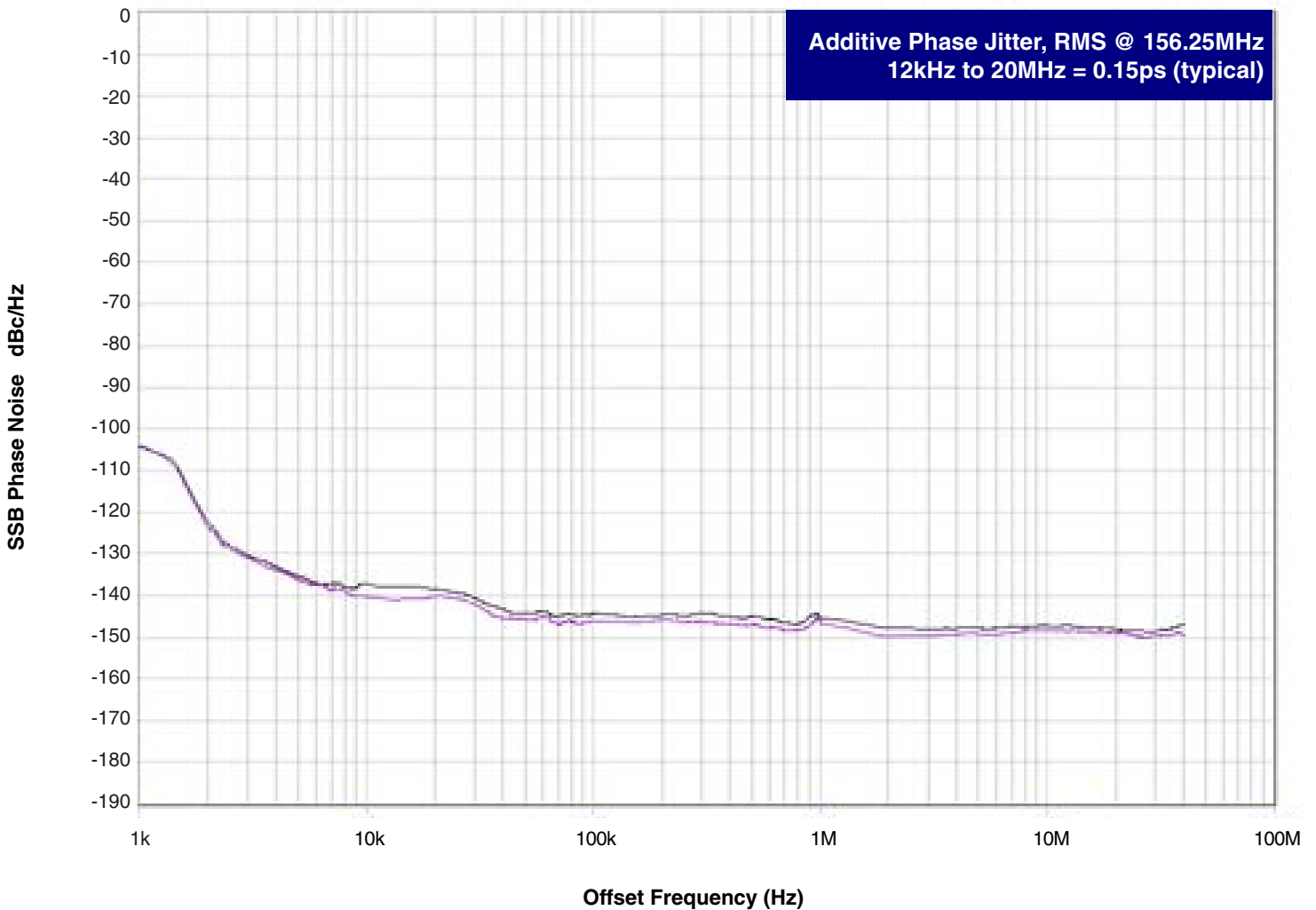
As with most timing specifications, phase noise measurements has issues relating to the limitations of the equipment. Often the noise floor of the equipment is higher than the noise floor of the device. This is illustrated above. The device meets the noise floor of what is shown, but can actually be lower. The phase noise is dependent on the input source and measurement equipment.

Parameter Measurement Information
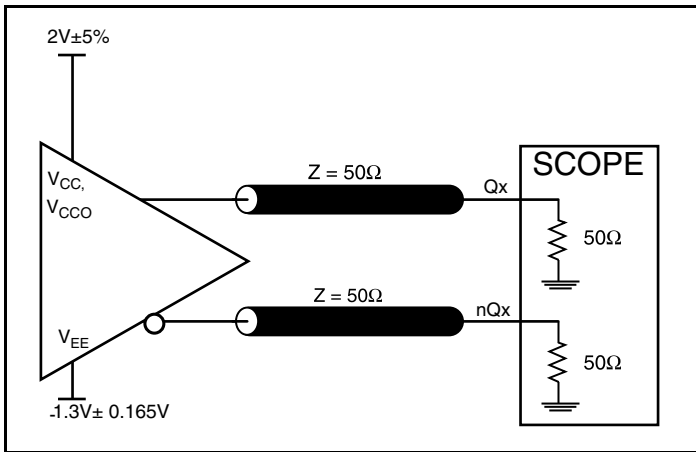
3.3V LVPECL Output Load AC Test Circuit
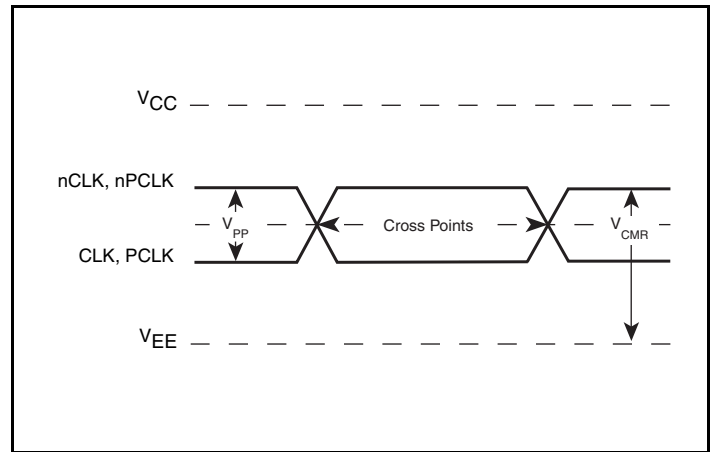
Differential Input Level
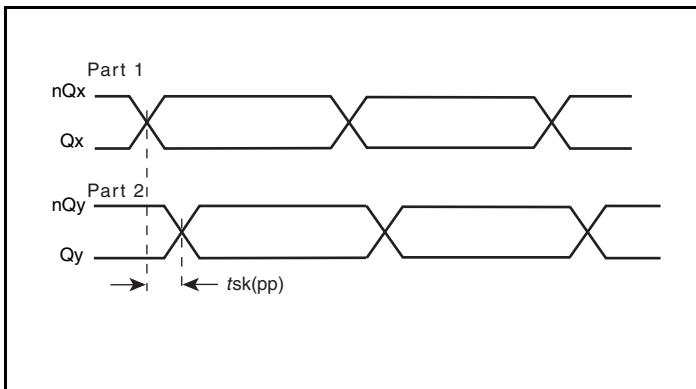
Part-to-Part Skew
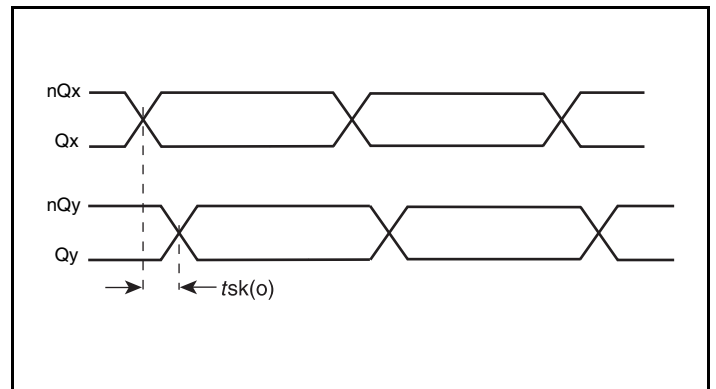
Output Skew
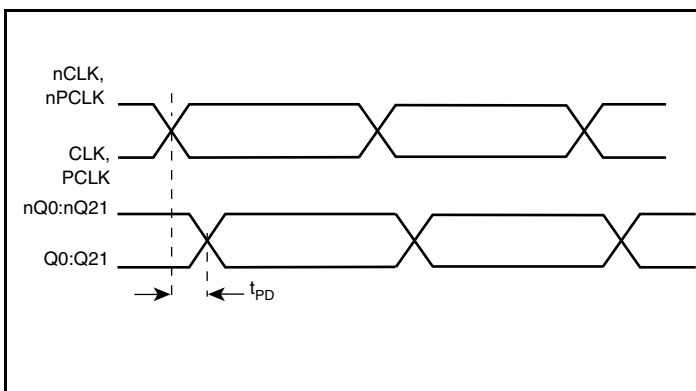
Propagation Delay
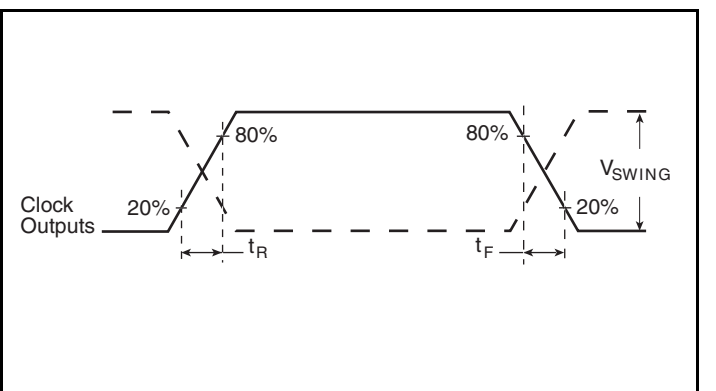
Output Rise/Fall Time

Parameter Measurement Information, continued

nQ0:nQ21

Q0:Q21

$$odc = \frac{t_{PW}}{t_{PERIOD}} \times 100\%$$
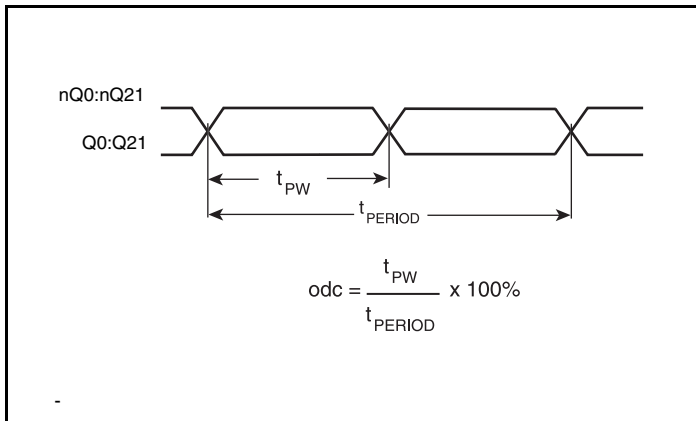
Output Duty Cycle/Pulse Width/Period

Applications Information

Recommendations for Unused Input and Output Pins

Inputs:

CLK/nCLK Inputs

For applications not requiring the use of the differential input, both CLK and nCLK can be left floating. Though not required, but for additional protection, a 1kΩ resistor can be tied from CLK to ground.

PCLK/nPCLK Inputs

For applications not requiring the use of the differential input, both PCLK and nPCLK can be left floating. Though not required, but for additional protection, a 1kΩ resistor can be tied from PCLK to ground.

LVCMOS Control Pins

All control pins have internal pull-ups or pull-downs; additional resistance is not required but can be added for additional protection. A 1kΩ resistor can be used.

Outputs:

LVPECL Outputs

All unused LVPECL outputs can be left floating. We recommend that there is no trace attached. Both sides of the differential output pair should either be left floating or terminated.

Wiring the Differential Input to Accept Single-Ended Levels

Figure 2 shows how a differential input can be wired to accept single ended levels. The reference voltage $V_{REF} = V_{CC}/2$ is generated by the bias resistors R1 and R2. The bypass capacitor (C1) is used to help filter noise on the DC bias. This bias circuit should be located as close to the input pin as possible. The ratio of R1 and R2 might need to be adjusted to position the V_{REF} in the center of the input voltage swing. For example, if the input clock swing is 2.5V and V_{CC} = 3.3V, R1 and R2 value should be adjusted to set V_{REF} at 1.25V. The values below are for when both the single ended swing and V_{CC} are at the same voltage. This configuration requires that the sum of the output impedance of the driver (Ro) and the series resistance (Rs) equals the transmission line impedance. In addition, matched termination at the input will attenuate the signal in half. This can be done in one of two ways. First, R3 and R4 in parallel should equal the transmission line impedance. For most 50Ω applications, R3 and R4 can be 100Ω. The values of the resistors can be increased to reduce the loading for slower and weaker LVCMOS driver. When using single-ended signaling, the noise rejection benefits of differential signaling are reduced. Even though the differential input can handle full rail LVCMOS signaling, it is recommended that the amplitude be reduced. The datasheet specifies a lower differential amplitude, however this only applies to differential signals. For single-ended applications, the swing can be larger, however V_{IL} cannot be less than -0.3V and V_{IH} cannot be more than V_{CC} + 0.3V. Though some of the recommended components might not be used, the pads should be placed in the layout. They can be utilized for debugging purposes. The datasheet specifications are characterized and guaranteed by using a differential signal.
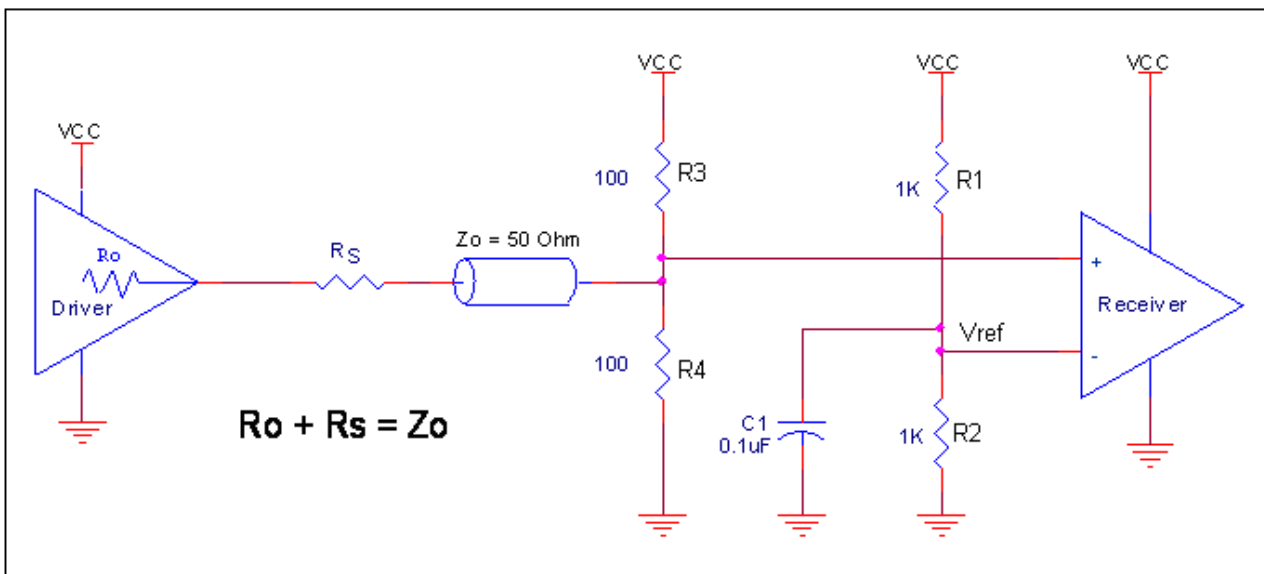

Figure 2. Recommended Schematic for Wiring a Differential Input to Accept Single-ended Levels

Differential Clock Input Interface

The CLK /nCLK accepts LVPECL, LVDS, HCSL, SSTL, LVHSTL and other differential signals. Both V_{SWING} and V_{OH} must meet the V_{PP} and V_{CMR} input requirements. *Figures 3A to 3F* show interface examples for the CLK/nCLK input driven by the most common driver types. The input interfaces suggested here are examples only. If the driver is from another vendor, use their termination recommendation. Please consult with the vendor of the driver component to confirm the driver termination requirements.
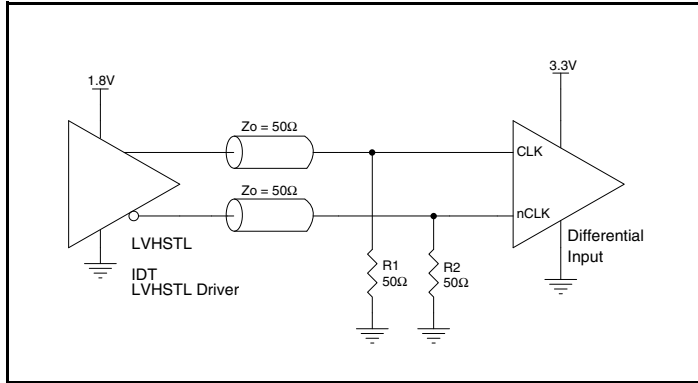
Figure 3A. CLK/nCLK Input Driven by an IDT Open Emitter LVHSTL Driver
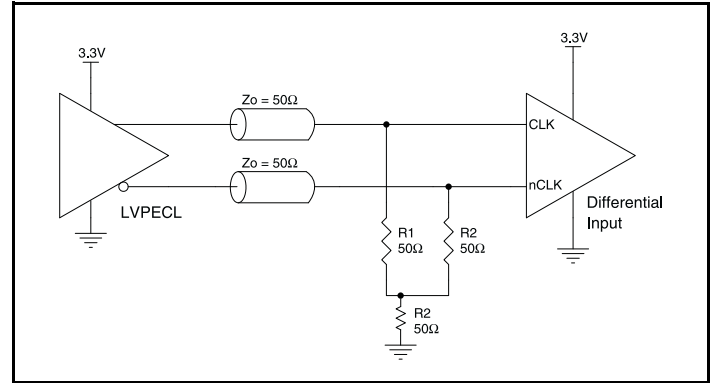
Figure 3B. CLK/nCLK Input Driven by a 3.3V LVPECL Driver
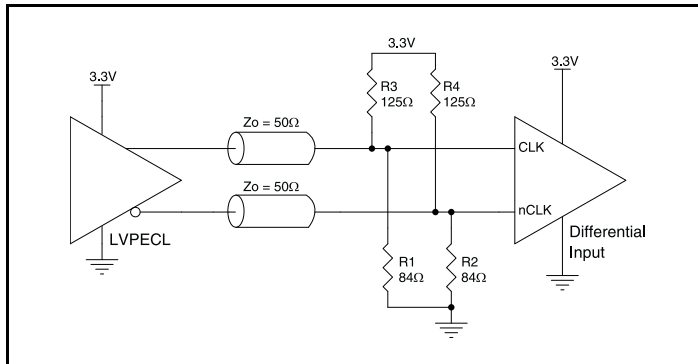
Figure 3C. CLK/nCLK Input Driven by a 3.3V LVPECL Driver
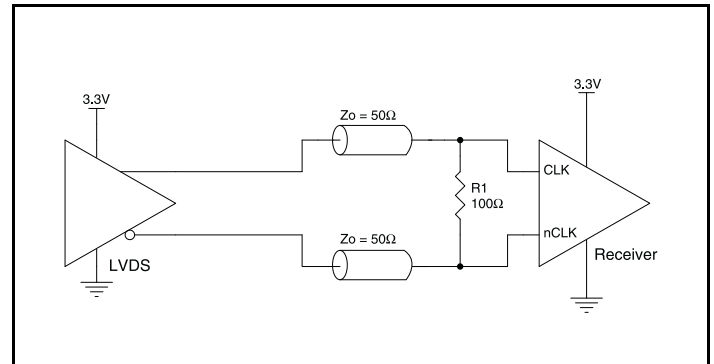
Figure 3D. CLK/nCLK Input Driven by a 3.3V LVDS Driver
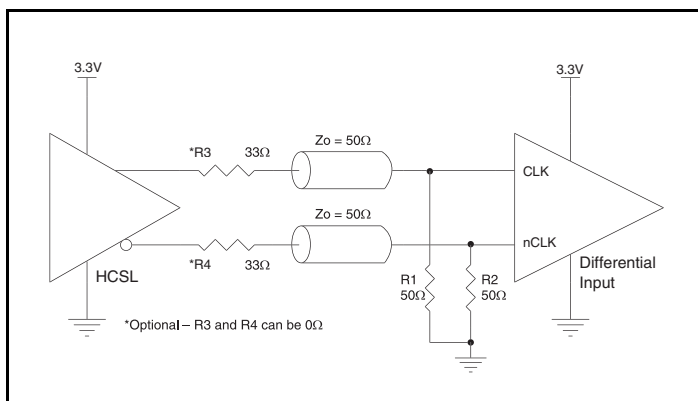
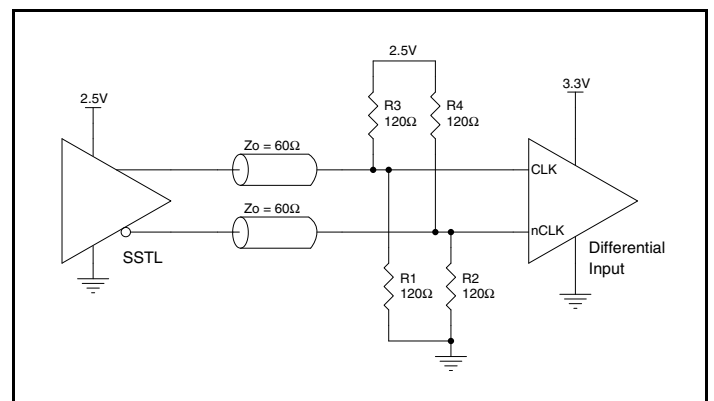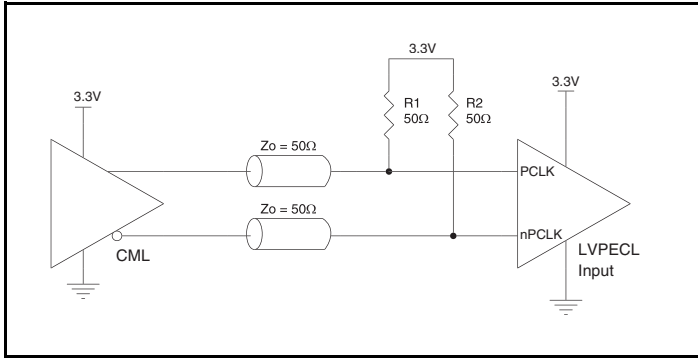**Figure 3E. CLK/nCLK Input Driven by a 3.3V HCSL Driver**

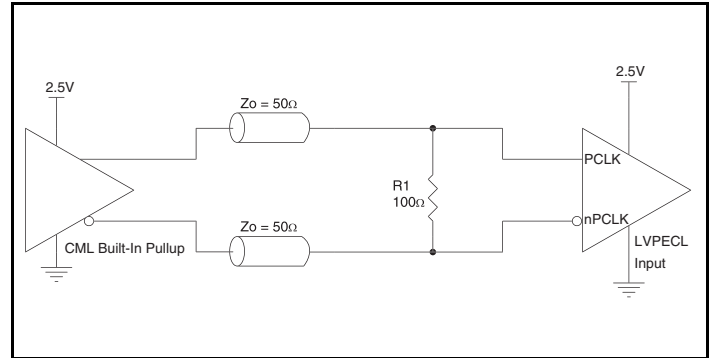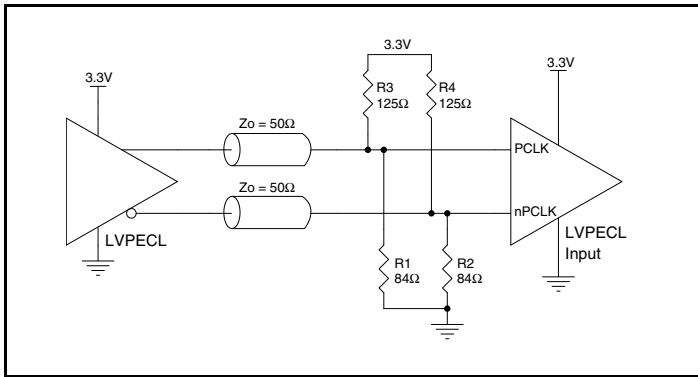**Figure 3F. CLK/nCLK Input Driven by a 2.5V SSTL Driver**

LVPECL Clock Input Interface

The PCLK /nPCLK accepts LVPECL, CML, SSTL and other differential signals. Both V_{SWING} and V_{OH} must meet the V_{PP} and V_{CMR} input requirements. *Figures 4A to 4E* show interface examples for the PCLK/nPCLK input driven by the most common driver types.

The input interfaces suggested here are examples only. If the driver is from another vendor, use their termination recommendation. Please consult with the vendor of the driver component to confirm the driver termination requirements.
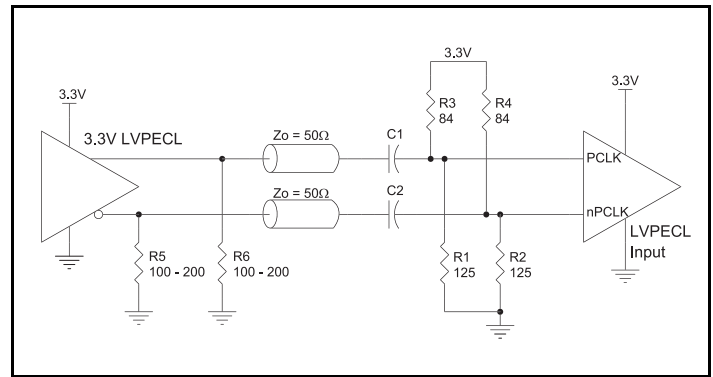
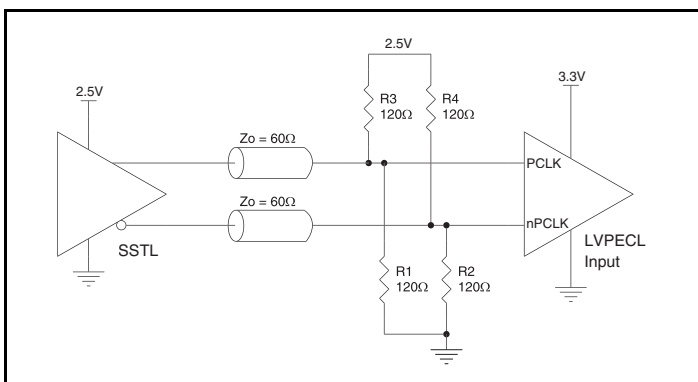**Figure 4A. PCLK/nPCLK Input Driven by an Open Collector CML Driver**

Figure 4B. PCLK/nPCLK Input Driven by a Built-In Pullup CML Driver

Figure 4C. PCLK/nPCLK Input Driven by a 3.3V LVPECL Driver

Figure 4D. PCLK/nPCLK Input Driven by a 3.3V LVPECL Driver with AC Couple

Figure 4E. PCLK/nPCLK Input Driven by an SSTL Driver

Termination for 3.3V LVPECL Outputs

The clock layout topology shown below is a typical termination for LVPECL outputs. The two different layouts mentioned are recommended only as guidelines.

The differential outputs are low impedance follower outputs that generate ECL/LVPECL compatible outputs. Therefore, terminating resistors (DC current path to ground) or current sources must be used for functionality. These outputs are designed to drive 50Ω

transmission lines. Matched impedance techniques should be used to maximize operating frequency and minimize signal distortion. *Figures 5A and 5B* show two different layouts which are recommended only as guidelines. Other suitable clock layouts may exist and it would be recommended that the board designers simulate to guarantee compatibility across all printed circuit and clock component process variations.
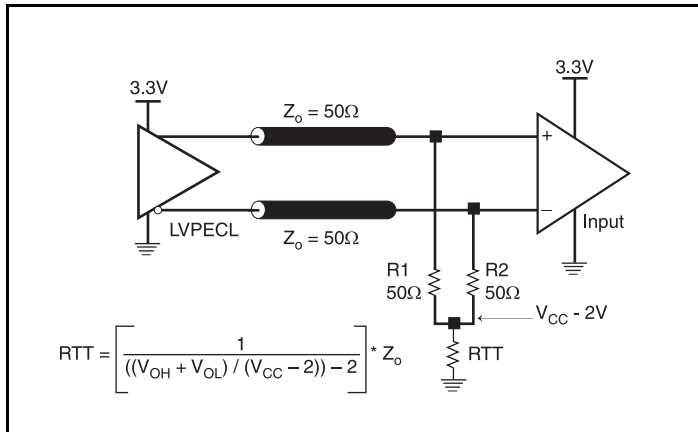

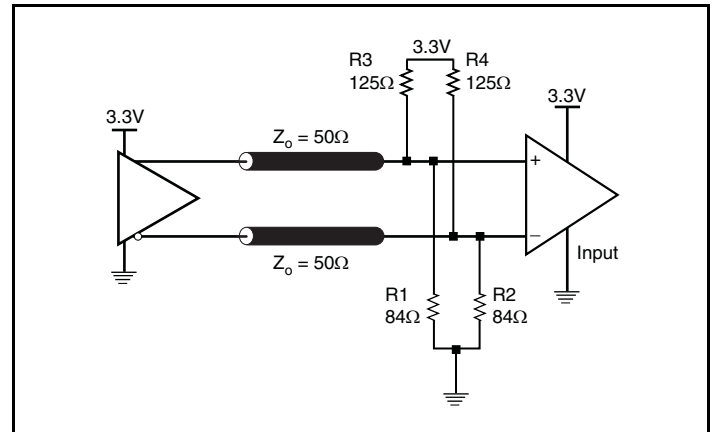**Figure 5A. 3.3V LVPECL Output Termination**

Figure 5B. 3.3V LVPECL Output Termination

EPAD Thermal Release Path

In order to maximize both the removal of heat from the package and the electrical performance, a land pattern must be incorporated on the Printed Circuit Board (PCB) within the footprint of the package corresponding to the exposed metal pad or exposed heat slug on the package, as shown in *Figure 6.* The solderable area on the PCB, as defined by the solder mask, should be at least the same size/shape as the exposed pad/slug area on the package to maximize the thermal/electrical performance. Sufficient clearance should be designed on the PCB between the outer edges of the land pattern and the inner edges of pad pattern for the leads to avoid any shorts.

While the land pattern on the PCB provides a means of heat transfer and electrical grounding from the package to the board through a solder joint, thermal vias are necessary to effectively conduct from the surface of the PCB to the ground plane(s). The land pattern must be connected to ground through these vias. The vias act as "heat pipes". The number of vias (i.e. "heat pipes") are application specific and dependent upon the package power dissipation as well as electrical conductivity requirements. Thus, thermal and electrical analysis and/or testing are recommended to determine the minimum number needed. Maximum thermal and electrical performance is achieved when an array of vias is incorporated in the land pattern. It is recommended to use as many vias connected to ground as possible. It is also recommended that the via diameter should be 12 to 13mils (0.30 to 0.33mm) with 1oz copper via barrel plating. This is desirable to avoid any solder wicking inside the via during the soldering process which may result in voids in solder between the exposed pad/slug and the thermal land. Precautions should be taken to eliminate any solder voids between the exposed heat slug and the land pattern. Note: These recommendations are to be used as a guideline only. For further information, refer to the Application Note on the *Surface Mount Assembly* of Amkor's Thermally/Electrically Enhance Leadframe Base Package, Amkor Technology.
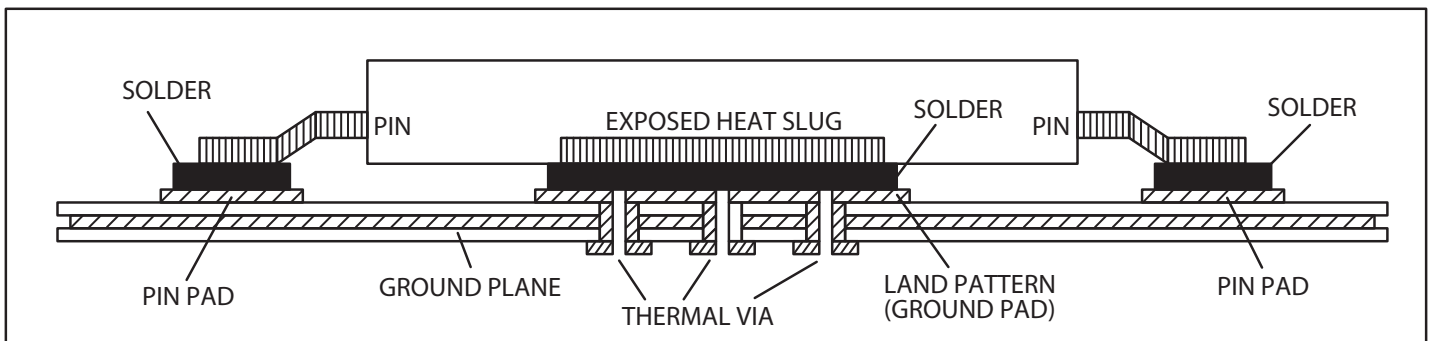
Figure 6. Assembly for Exposed Pad Thermal Release Path - Side View (drawing not to scale)

Power Considerations

This section provides information on power dissipation and junction temperature for the 8534-01.
Equations and example calculations are also provided.

1.   Power Dissipation.

The total power dissipation for the 8534-01 is the sum of the core power plus the power dissipated in the load(s).
The following is the power dissipation for V_{CC} = 3.465V, which gives worst case results.

NOTE: Please refer to Section 3 for details on calculating power dissipated in the load.

- Power (core)$_{MAX}$ = V_{CC_MAX} * I_{EE_MAX} = 3.465V * 230mA = **796.95mW**
- Power (outputs)$_{MAX}$ = **30mW/Loaded Output pair**
  If all outputs are loaded, the total power is 22 * 30mW = **660mW**

Total Power$_{\_MAX}$ (3.465V, with all outputs switching) = 796.95mW + 660mW = **1456.95mW**

2.   Junction Temperature.

Junction temperature, Tj, is the temperature at the junction of the bond wire and bond pad directly affects the reliability of the device. The maximum recommended junction temperature is 125°C. Limiting the internal transistor junction temperature, Tj, to 125°C ensures that the bond wire and bond pad temperature remains below 125°C.

The equation for Tj is as follows: Tj = θ_{JA} * Pd_total + T_A

Tj = Junction Temperature

θ_{JA} = Junction-to-Ambient Thermal Resistance

Pd_total = Total Device Power Dissipation (example calculation is in section 1 above)

T_A = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance θ_{JA} must be used. Assuming a moderate air flow of 200 linear feet per minute and a multi-layer board, the appropriate value is 17.2°C/W per Table 6 below.

Therefore, Tj for an ambient temperature of 85°C with all outputs switching is:

85°C + 1.457W * 17.2°C/W = 110.1°C. This is below the limit of 125°C.

This calculation is only an example. Tj will obviously vary depending on the number of loaded outputs, supply voltage, air flow and the type of board (multi-layer).

Table 6. Thermal Resistance θ_{JA} for 64 Lead TQFP, Forced Convection
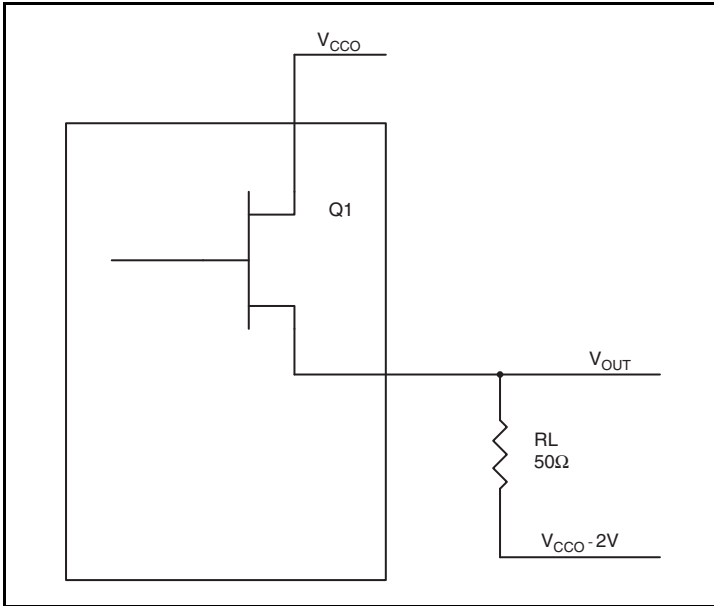
θ_{JA} **by Velocity**			
Linear Feet per Minute	**0**	**200**	**500**
Multi-Layer PCB, JEDEC Standard Test Boards	22.3°C/W	17.2°C/W	15.1°C/W

3. Calculations and Equations.

The purpose of this section is to calculate the power dissipation for the LVPECL output pair.

LVPECL output driver circuit and termination are shown in *Figure 7.*

Figure 7. LVPECL Driver Circuit and Termination

To calculate worst case power dissipation into the load, use the following equations which assume a 50Ω load, and a termination voltage of $V_{CCO} - 2V$.

- For logic high, $V_{OUT} = V_{OH\_MAX} = \mathbf{V_{CCO\_MAX} - 0.9V}$
  $(V_{CCO\_MAX} - V_{OH\_MAX}) = \mathbf{0.9V}$

- For logic low, $V_{OUT} = V_{OL\_MAX} = \mathbf{V_{CCO\_MAX} - 1.7V}$
  $(V_{CCO\_MAX} - V_{OL\_MAX}) = \mathbf{1.7V}$

Pd_H is power dissipation when the output drives high.

Pd_L is the power dissipation when the output drives low.

$Pd\_H = [(V_{OH\_MAX} - (V_{CCO\_MAX} - 2V))/R_L] * (V_{CCO\_MAX} - V_{OH\_MAX}) = [(2V - (V_{CCO\_MAX} - V_{OH\_MAX}))/R_L] * (V_{CCO\_MA} - V_{OH\_MAX}) = [(2V - 0.9V)/50\Omega] * 0.9V = \mathbf{19.8mW}$

$Pd\_L = [(V_{OL\_MA} - (V_{CCO\_MA} - 2V))/R_L] * (V_{CCO\_MAX} - V_{OL\_MAX}) = [(2V - (V_{CCO\_MAX} - V_{OL\_MAX}))/R_L] * (V_{CCO\_MA} - V_{OL\_MAX}) = [(2V - 1.7V)/50\Omega] * 1.7V = \mathbf{10.2mW}$

Total Power Dissipation per output pair = Pd_H + Pd_L = **30mW**

Reliability Information

Table 7. θ_{JA} **vs. Air Flow Table for a 64 Lead TQFP, E-Pad**

θ_{JA} vs. Air Flow			
Linear Feet per Minute	**0**	**200**	**500**
Multi-Layer PCB, JEDEC Standard Test Boards	22.3°C/W	17.2°C/W	15.1°C/W

Transistor Count

The transistor count for 8534-01 is: 1474

Package Outline and Package Dimensions

Package Outline - Y Suffix for 64 Lead TQFP, E-Pad

Table 8. Package Dimensions for 64 Lead TQFP, E-Pad

JEDEC Variation: ACD All Dimensions in Millimeters			
Symbol	**Minimum**	**Nominal**	**Maximum**
N		64	
A			1.20
A1	0.05	0.10	0.15
A2	0.95	1.00	1.05
b	0.17	0.22	0.27
c	0.09		0.20
D & E		12.00 Basic	
D1 & E1		10.00 Basic	
D2 & E2		7.50 Ref.	
D3 & E3	4.5	5.0	5.5
e		0.50 Basic	
L	0.45	0.60	0.75
θ	0°		7°
ccc			0.08

Reference Document: JEDEC Publication 95, MS-026

Ordering Information

Table 9. Ordering Information

Part/Order Number	Marking	Package	Shipping Packaging	Temperature
8534AY-01LF	ICS8534AY-01LF	64 Lead TQFP, Lead-Free	Tray	0°C to +85°C
8534AY-01LFT	ICS8534AY-01LF	64 Lead TQFP, Lead-Free	Tape & Reel	0°C to +85°C

Revision History Sheet

Rev	Table	Page	Description of Change	Date
A		15	Updated Package Outline and Package Dimensions.	11/19/04
A	T9	1 12 13 18	Features Section - added lead-free bullet. Added *Recommendations for Unused Input and Output Pins* section. Updated *EPad Thermal Release Path* section. Ordering Information Table. Added lead-free part number, marking and note. Updated format throughout the datasheet.	12/06/07
A		1	Pin Assignment - rotated pin orientation 90° back to original orientation, (datasheet publication dated December 6, 2007).	5/09/08
B	T1 T4C T6	1 2 5 6 10 11 12	Features section, change Additive Phase Jitter spec to 0.15ps typical. Pin Description Table - corrected nQ0, Q0 numbering to 62, 63. LVPECL DC Characteristics Table - corrected notes. AC Characteristics Table - corrected Additive Phase Jitter spec to 0.15ps typical. Updated *Wiring the Differential Input to Accept Single-ended Levels* application note. Corrected Differential Clock Input Interface Application note. Corrected LVPECL Clock Input Interface Application note. Updated to new header/footer format.	3/28/11
C	T9	1 19	Features section - updated package bullet. Ordering Information Table - deleted "Lead-Free" in Package column for non-LF parts. Corrected lead-free Shipping Packaging information. Updated header/footer to new format.	12/1/15

Corporate Headquarters
6024 Silver Creek Valley Road
San Jose, CA 95138 USA
www.IDT.com

Sales
1-800-345-7015 or 408-284-8200
Fax: 408-284-2775
www.IDT.com/go/sales

Tech Support
www.idt.com/go/support